Development of an Adiabatic RF Neutron Spin Flipper at the China Spallation Neutron Source*

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A new prototype adiabatic RF-flipper was recently developed at the China Spallation Neutron Source. The prototype device was calibrated at the test beamline BL-20 over a wavelength range of 0.6 Å to 5.5 Å, and achieved a flipping efficiency of 97% for neutron wavelengths above 4 Å. During the development of the adiabatic RF-flipper, finite element method and spin transformation simulations were applied to precisely determine the magnetic field configuration and neutron spin-flip efficiency. This work demonstrates the design and optimization of the adiabatic RF-flipper for a specific neutron beamline, where the dependence of the flipping efficiency on neutron wavelength can be analyzed through simulation and numerical calculation for pulsed neutron beams.

Keywords: neutron spin-flipper, polarized neutron, spallation neutron source

I. INTRODUCTION

Neutron scattering is a widely used method for physics and material research. In recent years, several large-scale neutron facilities have been developed across the world to adsvance neutron scattering techniques and applications. Existing neutron facilities have also undergone continuous upgrades to allow for many cutting-edge scientific measurements. Among neutron scattering experimental methods, polarized neutrons play a key role and have a wide range of applications. This is especially true for weak magnetic scattering, where advanced neutron polarization instruments are usually required for effective measurements [1]. Instruments that can utilize polarized neutrons include Small-Angle Neutron Scattering (SANS), neutron reflectometry, neutron imaging, neutron diffraction, inelastic neutron scattering and Neutron Spin-Echo (NSE) [2–12].

With the support of polarized neutron techniques, many new experimental methods have emerged that contribute to research in various fields [13–20]. Notable techniques in- clude longitudinal polarization analysis, XYZ polarization analysis, and Spherical Neutron Polarimetry (SNP). Longitudinal polarization analysis involves performing neutron spin

fluxes to pass through without any neutron scattering or ab-

55 sorbing. It is used as a key piece of equipment in high-

56 resolution material dynamics studies in NSE, reducing back-

57 ground signals in neutron scattering experiments, and in other

58 experimental techniques using wide wavelength spectrum po-

23 analysis along a one-dimensional direction in experiments, al-24 lowing for the measurement and analysis of non-spin-flip and 25 spin-flip cross-sections under specific scattering geometries,

where the applied magnetic field (guide field) is often applied

to adjust the polarization of the incident neutron beam, pro-

viding magnetic information about the materials. XYZ po-

29 larization analysis is suitable for simultaneous measurements

30 in a multidetector polarization analysis experiment, but it still

31 poses difficulties in measuring the spin rotation quantities re-

32 lated to chiral magnetic scattering. The SNP establishes a

33 zero magnetic field region to shield the magnetic field, en-

34 abling precise general polarization analysis over a wide scat-

35 tering angle range. For all mentioned techniques, controlling

36 the neutron spin state during the measurement is an essen-

37 tial step. The neutron spin flipper (NSF) is widely used in

³⁸ polarized neutron experiments due to its capability to control 39 the neutron spin state [21-23]. In modern neutron experi-40 ments, using a spin flipper with high transmission and excel-41 lent flipping efficiency is crucial because it can improve the 42 experiment's precision and simplify the measurement proce-43 dure and data analysis. Over the past decades, various types of spin flippers were 45 developed to satisfy the demands of different neutron sources. 46 These include Mezei flippers [10], Drabkin spin flippers 47 [24], thin film flippers [25], resonant RF-flippers [26], 48 superconductor-based flippers [16, 27-29], and adiabatic RF-49 flippers [30, 31]. Due to their nature of controlled preces-50 sion, the Mezei flipper and thin film flippers are limited to 51 one specific neutron wavelength at a time. Whereas an adi-52 abatic RF-flipper can simultaneously support a wide range of neutron wavelengths and allow neutron beams with larger

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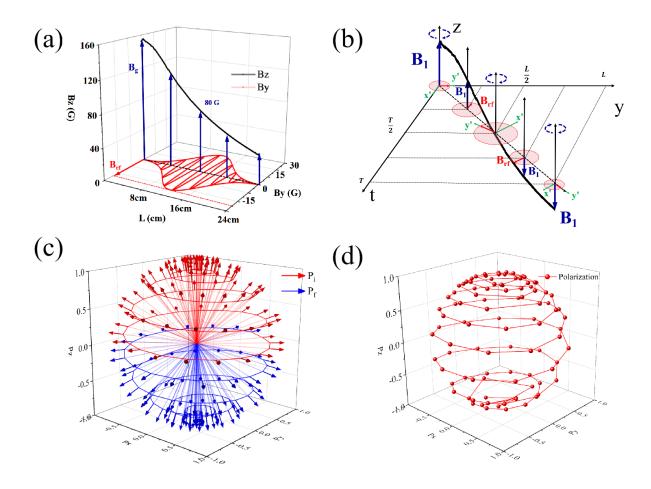


Fig. 1. (a) The laboratory frame is applied in this design. The direction of the neutron beam and the RF-field (\mathbf{B}_{RF}) are parallel to the y-axis, and the gradient field ($\mathbf{B}_1 = \mathbf{B}_g - \mathbf{B}_0$) direction is set as the z-axis. l is the length of the flipper. (b) The rotating frame rotates uniformly clockwise along the z-axis with an angular velocity ω . In the rotating frame, the RF-field component (rotating with the frame) is parallel to the y'-axis, the gradient field \mathbf{B}_g is parallel to the z-axis, and the RF-field component rotating at a different frequency is neglected. (c-d) The orientation of the neutron polarization in the adiabatic RF-flipping process. (c) Theoretical progression of a 4 Å neutron polarization vector within an ideal linear field (from 130 to 70 G). (d) Calculated progression of a 4 Å neutron polarization vector for the designed adiabatic RF-flipper fields \mathbf{B}_g and \mathbf{B}_{RF} shown in Fig. 1.

larized neutrons. With the development of modern neutron sources, time-of-flight based neutron instruments at pulsed neutron sources require more detailed design, such as Spallation Neutron Source (SNS) at Oak Ridge National Laboratory (ORNL), Material and Life Science Experimental Facility at Japan Proton accelerator Research Complex (J-Parc) and the China Spallation Neutron Source (CSNS). Especially for newly developed beamlines, it is advantageous to consider polarization capability during the design and construction process, so that optics and magnetic field configurations can be collectively optimized.

For a typical time-of-flight neutron beamline, the combination of a polarizing supermirror and an adiabatic RF-flipper is one of the few solutions to generate polarization and control the spin state for a wide-range of neutron wavelengths with high polarization and transmission efficiency. The polarizing supermirror and flipper combination usually serves as a sub-

⁵⁹ larized neutrons. With the development of modern neutron ⁷⁶ stitutional part of the neutron guide, and hence contribute to ⁷⁷ the overall neutron beam optics. Two options exist for de⁸⁰ neutron sources require more detailed design, such as Spal⁸⁰ ploying the polarizing supermirror and adiabatic RF-flipper.

One option is to place the two components at the beginning of the neutron guide section where the neutron beam has lower cross-sectional area and divergence, so that undesired reflections from the supermirror can be avoided. For such a configuration, the neutron polarization must be transferred to the sample position by a relatively long guide field system, and the adiabatic RF-flipper must be incorporated with the neutron guide across. Typical beamlines such as the GP-SANS at ORNL, LET cold neutron multi-chopper spectrometer, and Larmor SANS/SESANS beamline at the ISIS neutron and Muon Source (UK) adopt such a design during commissioning. The other option is to install the polarizing supermirror and adiabatic RF-flipper within close approximate to the sample, usually inside the scattering room for easy access

94 ally require more sophisticated control of the magnetic field 149 transformed as follows: 95 design because the sample stage and magnetic field environ-96 ment affect the performance of the adiabatic RF-flipper and 150 polarization transfer. In general, this configuration is less restrictive on the neutron beam optics but would still require 151 The neutron polarization undergoes an adiabatic change from 99 tubes for better transmission. 100

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In this study, we detail the development of the first in-house manufactured adiabatic RF-flipper at the CSNS which provides a platform for conducting scientific research in many frontier disciplines [32–34]. The prototype device is tested with a time-of-flight polarized neutron beam for performance calibration. The optimization process of the device's key parameters is also explained, along with a comparison between the simulated results and experiment data. The results show that with numerical simulations precise control of the neutron polarization can be predicted, thereby facilitating customization of future adiabatic RF-flippers at the CSNS for different 112 polarized neutron experiments.

PRINCIPLES OF RF-FLIPPING

115 (AFP) are well established [35], accomplishing spin flips 168 formed into a rotating frame, as illustrated in Fig. 1(b), and through a helical spinning process achieved by a combina- 169 the adiabaticity of the spin flip transformation in the rotating 117 tion of gradient fields and a corresponding radio-frequency 170 frame is examined using Eq. (2). This process is repeated 118 field (RF-field). The two fields are set perpendicular to each 171 for multiple iterations of the magnetic field design, and the other along the neutron path, as shown in Fig. 1(a), with the 172 optimized design is fabricated into a prototype for characdirection of the gradient field \mathbf{B}_q inside the flipper along the 173 terization tests and neutron experiments. Further analysis of distance (l) from the entrance into the RF-flipper region, il- 177 tron beamline. This comparison provides insights into the 124 125 lustrating their spatial variations.

ponents $B_0 + B_1$, with B_0 as the constant "center field" that 180 Additionally, the research demonstrated the resilience of the defines the frequency of B_{RF} as $\omega_{RF}=-\gamma_n B_0$, where γ_n is 181 prototype device to external magnetic field influences, which the neutron gyromagnetic ratio. The varying component \mathbf{B}_1 182 is discussed further in the manuscript. decreases from positive to negative along the neutron flight path. The RF-field \mathbf{B}_{RF} is generated by a solenoid with the defined frequency ω_{RF} , and its maximum magnitude aligns at the center position where $\mathbf{B}_q = \mathbf{B}_0$, so that the adiabatic RF-flipping condition is achieved.

ization undergoes a flipping process, which is commonly de- 186 nents, shown in Fig. 2, where 1. Enclosure of the device. scribed in a rotating frame about the z-axis with a frequency 187 2. Iron plates and permanent magnets generating gradient of ω_{RF} as shown in Fig. 1(b). Within the rotating frame, po- 188 magnetic field. 3. Detection coil. 4. Solenoid generatlarization along the z-axis is consistent with the laboratory 189 ing the RF-field. 5. Front and 6. Back guide fields comframe, while the effective magnetic field along the z-axis ${f B}_g$ 190 posed of iron plates and permanent magnets. The enclois reduced by ${\bf B}_0$ so that only ${\bf B}_1$ remains. In the rotating 191 sure of the RF-flipper is composed of several non-magnetic frame, the RF-field component $(\frac{1}{2}B_{RF}(l))$ rotating with the 192 bakelite pieces, which ensure the stability of the magnetic frame is parallel to the y'-axis, the gradient field B_1 is par- 193 field inside the flipper. The gradient field in the z-direction 144 allel to the z-axis, and the RF-field component rotating at 194 is generated by symmetrically installed iron plates and per- $_{145}~\omega=-2\omega_{
m RF}$ is neglected. The RF-field ${f B}_{
m RF}$ is decomposed $_{195}$ manent magnets. The central iron plates are tilted by ± 25 into a $\frac{1}{2}|B_{RF}(l)|$ component parallel to the y'-axis and a fast 196 degrees from the horizontal plane. The iron plates have di-147 counter-rotating component that is usually neglected. The ef- 197 mensions of $28 \text{ cm} \times 10 \text{ cm} \times 0.2 \text{ cm}$. Six Nd₂Fe₁₄B magnets

 $_{93}$ and experiment flexibility. However, such configurations usu- $_{148}$ fective magnetic field \mathbf{B}_e experienced by the neutron is then

$$\mathbf{B}_e = B_1(l)\hat{\mathbf{z}}' + \frac{1}{2}B_{RF}(l)\hat{\mathbf{y}}'$$
 (1)

the adiabatic RF-flipper to be compatible with neutron flight 152 + $\hat{\mathbf{z}}'$ into + $\hat{\mathbf{y}}'$ and then $-\hat{\mathbf{z}}'$ within the rotating frame, as long as the adiabatic condition in Eq. (2) is satisfied:

$$\left| \frac{d}{dt} \left(B_1(l) + \frac{1}{2} B_{RF}(l) \right) \right| \ll 2\pi \gamma_n \left| B_1(l) + \frac{1}{2} B_{RF}(l) \right|^2$$
(2)

The adiabatic transition in the rotating frame becomes a 156 helical trajectory when transformed back to the laboratory 157 frame, as shown in Fig. 1(c). The initial polarization vector 158 P_i moves towards the xy-plane along red helical lines, and then aligns with the z-axis in the opposite direction as shown 160 by the blue helical lines. The final polarization vector \mathbf{P}_f is 161 antiparallel to the z-axis.

The initial step in designing the prototype adiabatic RF-163 flipper involved tailoring the gradient magnetic field to the anticipated magnetic field environment. The magnetic field in the RF-flipper is presented in Fig. 1(a), and the calculated 166 evolution of the polarization vector is depicted in Fig. 1(d). The principles of spin flipping by adiabatic fast passage 167 To assess the RF-flipping process, the magnetic field is transdirection in the laboratory frame. The RF-field (B_{RF}) is 174 the neutron polarization progression is conducted with meaparallel to the neutron flight path and along the y-direction. 175 surements of the prototype's magnetic field, and this is com-The magnitudes of both fields are plotted with respect to the 176 pared to the neutron flipping efficiency measured with a neu-178 flipping performance and design accuracy across a range of The gradient field ${f B}_g$ can be decomposed into two com- 179 neutron wavelengths, which will guide future optimization.

III. MAIN COMPONENTS DESIGN

The RF-flipper prototype is designed based on the princi-In the adiabatic RF-flipping condition, the neutron polar- 185 ples previously discussed and consists of six main compo-

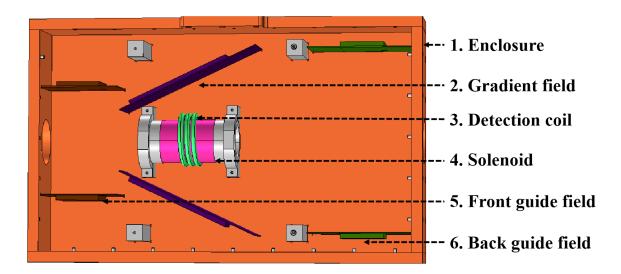


Fig. 2. Cross-sectional schematic view of the adiabatic RF-flipper prototype.

199 erate a gradient field ranging from 160 G to 40 G along the 236 generate sufficient magnetic field at ω_{RF} frequency when its 205 $15\,\mathrm{cm} \times 8\,\mathrm{cm} \times 0.2\,\mathrm{cm}$. 206

The solenoid is a single-layer coil 10.25 cm long and 8 cm 244 $_{208}$ in diameter with 93 turns of 1.1 mm enamel-coated copper $_{245}$ mized for the design with a $10.00\,\Omega$ resistor and a $2.05\,\Omega$ 210 strate. The inductance of the coil should ideally be 0.54 mH 247 ity factor of 48.94. Fig. 3(a) illustrates the RLC circuit dia-211 and is measured at 0.40 mH with an RLC digital bridge. The 248 gram for the developed system, where V represents the power 212 solenoid is fixed between two tilted iron plates, generating 249 source composed of a standard function generator and RF-213 the RF-field \mathbf{B}_{RF} . The solenoid creates an effective RF field 250 amplifier, R is the total resistance of the circuit, C is the total 214 region approximately 24 cm long between the gradient field 251 capacitance, and L is the inductance of the solenoid. Using 215 generating iron plates and can accommodate a 6 cm diameter 252 multiple resistors and capacitors in an RLC circuit can in-216 neutron beam passing through the spin flipper. The detec- 253 crease the heat dissipation area and improve the thermal station coil consists of three rounds of wire wrapped around the 254 bility of the circuit. solenoid, which can be moved along the solenoid to moni- 255 tor the generated RF-field at different locations. The solenoid 256 70 W, and the result is shown in Fig. 3(c-e). In Fig. 3(c), when 220 function generator and amplified by a 70 W RF-amplifier. 221

into the solenoid. The components of the matching box in- 261 clude a protective casing, a cooling fan, and resistive and 262 capacitive components of the RLC circuit. The external solenoid of the RF-flipper creates the inductance of the RLC circuit. The resonant frequency (f) of the RLC circuit is set to $_{265}$ RLC resonance frequency is $1.44\,\mathrm{kHz}$ above the design frequency be consistent with the neutron Larmor precession frequency $\omega_{\rm RF}=233.26\,{\rm kHz},$ and the bandwidth exceeds the defined by B_0 and consequently leads to a capacitance of a_{267} design value by $0.03\,\mathrm{kHz}$. These minor discrepancies are 1150 pF for the match-box.

contact resistance/capacitance, the resonant frequency of the 270 ing prototype functions sufficiently well to cover the neutron 234 RLC-circuit may shift and result in a reduced power output at 272 RF-flipping conditions.

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198 (4 cm \times 2 cm \times 1 cm) are attached to the iron plates to gen- 235 the set frequency of ω_{RF} . The solenoid circuit is designed to neutron flight path. This gradient field is nonlinear, and the 237 total output is above $1/\sqrt{2}$ times the maximum power, thus center field $B_0 = 80 \,\mathrm{G}$ is generated near the geometrical cen- 238 allowing for a flexible "bandwidth" of the RLC-circuit rester of the iron plates. Guide fields are installed both upstream 239 onance frequency. This is achieved by introducing a proper and downstream of the central magnetic field components to 240 resistance load into the match-box circuit. The resistors affect isolate external stray magnetic fields. Each guide field com- 241 the quality factor (Q) of the RLC circuit, and the dependence ponent consists of a pair of iron plates with dimensions of 242 between Q and R, as well as the corresponding bandwidth, is 243 shown in Fig. 3(b).

The RLC circuit bandwidth and output power are optiwire wound onto a plastic-steel (Polyvinyl Chloride) sub- 246 solenoid, resulting in a bandwidth of 4.79 kHz and a qual-

The RLC circuit is then tested with a total power up to is powered by a sine-wave current generated from a standard 257 the RF-power of the sine-wave signal is input into the RLC 258 circuit, a stable sine-signal can be detected by the detection A matching circuit (match-box) tunes the RLC circuit of 259 coil using Faraday's electromagnetic induction law, which inthe solenoid to ensure sufficient RF-power can be forwarded 260 dicates that the designed RLC circuit generates an RF-field with the same waveform and frequency as the source signal.

The strength of the RF-field monitored by the detection coil $_{\rm 263}\,$ peaks around $234.70\,{\rm kHz}$ with a bandwidth of $4.82\,{\rm kHz}$ (from 264 232.26 kHz to 237.08 kHz), as shown in Fig. 3(d). The tested 268 caused by the BNC cable and solenoid connecting wire not Due to the influence of thermal perturbations and wire- 269 being included in the design process. However, the result-

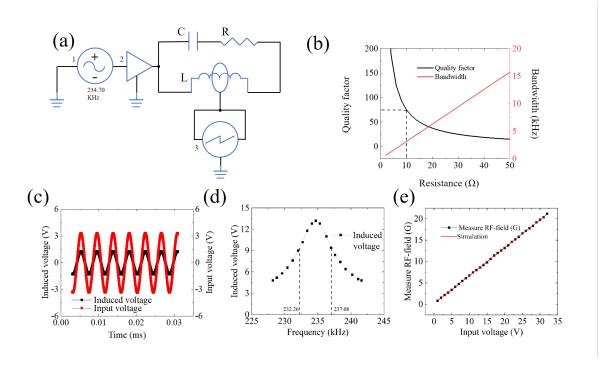


Fig. 3. (a) Schematic diagram of the RLC circuit and detection circuit. 1, 2, and 3 are respectively the function generator, RF-amplifier, and oscilloscope. (b) The resistance dependence of Q (black line, left axis) and bandwidth (red line right axis). (c) The red line is the waveform signal generated from the function generator, and the black line is the induced signal from the detection coil. (d) The frequency dependence of the RF-field measured by the induced signal in the detection coil. (e) The input voltage dependence of the RF-field measured with the detection coil.

Fig. 3(e) shows the induced magnetic field strength mea- 300 pacitance parameters may lead to a slight decrease in the RF 273 sured by the detection coil as the input voltage increases. The 302 field generated by the circuit. slope obtained after performing a linear fit on the dependency between the RF field and the input voltage is 0.65, indicating that the input power of the solenoid is stable and not significantly affected by voltage fluctuations. By combining the magnetic field formula of the energized solenoid with Faraday's law of electromagnetic induction, the slope can be calculated based on the parameters of the solenoid and detection 305 298 makes it susceptible due to measurement position. Thirdly, as 322 a guiding field inside the flipper aims to reduce external mag-

SIMULATIONS OF MAGNETIC FIELDS AND NEUTRON POLARIZATION

The gradient fields generated by the main components are coil. The magnetic field inside the solenoid is generated by 306 simulated using the three-dimensional finite element method the input voltage and can be approximated with $B=\mu_0\frac{NI}{L_0}$, so in COMSOL Multiphysics software to establish a magnetic where μ_0 is the permeability of air, N is the number of turns so field model of the device [36]. In the simulated model, the in the solenoid, L_0 is the length of the solenoid, and I is 309 relative permeability of the iron plate is 6400, which is calthe current flowing through the solenoid. The relationship 310 culated based on the hysteresis loop of the actual iron plate. between the RF-field and the induced electromotive force is 311 The material of the permanent magnet is N35, and its residual $E = -n\frac{d\Phi}{dt}$, where E is the induced electromotive force in 312 magnetic flux density modulus reaches as high as 1.2T. Usthe detection coil, n is the number of turns in the detection 313 ing the magnetic field interface in the AC/DC module, the incoil, and Φ is the magnetic flux on the cross-section of the 314 duced magnetic field generated by the permanent magnets is detection coil. The calculated ratio between the measured RF 315 calculated at different angles based on Ampere's Law. In the magnetic field and input voltage is 0.71, while the fitted slope 316 region of the solenoid where polarized neutrons pass through, obtained from measurements is 0.65. There are three main 317 the extremely fine mesh that is the default setting in COMSOL reasons for this deviation. Firstly, the solenoid has a finite 318 is adopted, while in other regions, the larger extra fine mesh length, resulting in a center magnetic field that is only approx- 319 is used. This method of mesh division aims to ensure compuimated equal to that of an infinite solenoid. Secondly, the de- 320 tational efficiency while maintaining sufficient mesh density tection coil is wound around the center of the solenoid, which 321 to improve simulation accuracy. The design feature of adding 299 the operating time increases, variations in resistance and ca- 323 netic interference and enhance its stability. Therefore, this 324 model does not include the guiding magnetic fields outside

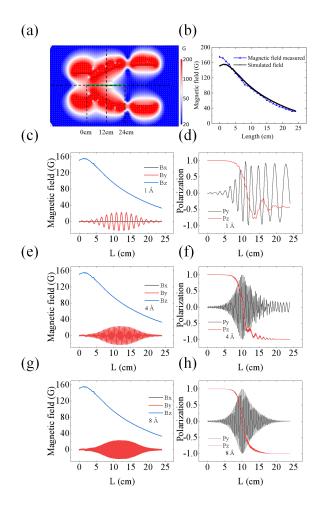


Fig. 4. (a) Cross-sectional view of the flux distribution in the neutron path. The white arrows are the directions of the magnetic field lines. The horizontal green line is the center line of the RF-field region generated by the 10 cm long solenoid. (b) The black line represents the one-dimensional vertical magnetic field distribution at the specific position indicated by the green line in (a). The magnetic field traversed by neutrons of (c) 1 Å, (e) 4 Å, and (g) 8 Å in the laboratory frame. The blue line represents the actual onedimensional vertical magnetic field distribution measured along the center axis of the solenoid, with the starting point of L at 0 cm in (a). The red line represents the RF-field experienced by neutrons of different wavelengths in (c), (e) and (g). The RF-field has a position dependent profile with a maximum value of 22 G at the center of the solenoid, corresponding to the measured RF-field at the maximum 355 input power in Fig. 3(e). The gradient magnetic field (blue line) is 356 the simulated field from (b). The neutron polarization calculated for 357 neutron wavelengths of (d) 1 Å, (f) 4 Å, and (h) 8 Å.

the flipper enclosure or the fields located before and after it, 361 but it does include the guiding magnetic fields at the front and 362 back inside the flipper enclosure. 327

Fig. 4(a) shows a plot of a cross-section of the magnetic 364 side is about 1 G lower than the simulated one. 328 field through the neutron path. Moving along the neutron 365 path from left to right, the magnetic field gradually increases 366 the detection coil at different positions inside the solenoid. to 160 G at the entrance of the spin flipper before gradually 367 When the flipper is activated, both the magnitude and fredecreases across the 24 cm long RF-field region. Fig. 4(b) 368 quency of the RF field remain constant, but the neutron timeshows that the simulated gradient magnetic field within the 369 of-flight through the RF-field varies directly withits wave-

334 RF-field region varies from 160 G to 40 G, with a magnetic field strength of 80 G at the center position.

An algorithm developed for tracking neutron spin procession is used to numerically solve the Bloch equations to calculate the spin change of the polarized neutrons in the magnetic field and to verify the RF-flipper's efficiency for neutrons of different wavelengths [37]. When using this method to solve the Bloch equations, it is necessary to set the parameters including neutron wavelength, magnitudes of magnetic fields in the xyz directions, step size, and direction of the neutron's initial polarization vector. The step size is defined by the spatial distance between two magnetic field vectors. When the magnetic field changes over time, the step size needs to be converted based on the neutron flight speed to ensure that the magnetic field conditions remain consistent for neutrons of different wavelengths. The change in the neutron polarization vector, adiabatic parameter, and angle between polarization vector and magnetic field can be obtained at several wavelengths with different magnetic field parameters by using this 353 model.

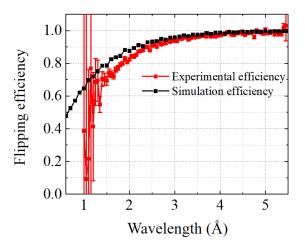


Fig. 5. The simulated and experimental flipping efficiency of the adiabatic RF-flipper for 1.0-5.5 Å polarized neutrons.

The measured gradient field B_g shown in Fig. 4(b) is obtained by performing Gaussmeter measurements at uniform positions along the path represented by the green line in Fig. 4(a), followed by Gaussian fitting. In the central region of the solenoid, the actual measured magnetic field is consistent with the simulated magnetic field in terms of size and trend. However, at both ends of the solenoid, there are differences between the actual and simulated magnetic fields. The actual magnetic field on the left side is about 20 G higher than the simulated one, while the actual magnetic field on the right

The amplitude of the RF-field (B_{RF}) is measured with

371 more magnetic field oscillations the neutron experiences. 428 noticeable angle to the z-axis. Upon exiting the RF-flipper, 372 Consequently, in the simulation, the position-dependent RF- 429 a significant component of the polarization vector remains 373 field experienced by the neutron changes with the neutron 430 within the xy-plane. This indicates that the RF flipper is not wavelength such that $B_{\rm RF}(l)=B_{\rm RF}\sin\left(\frac{B_0\gamma_{\rm n}lm\lambda}{h}\right)$, where ⁴³¹ fully effective in flipping 1 Å neutrons. ${}_{375}$ ${\bf B}_{\rm RF}$ as shown by the red curve in Fig. 4(c)(e)(g), m is the $_{376}$ rest mass of a neutron, λ is the neutron wavelength, and $_{377}$ h is Planck's constant. Expressing the RF-field as a func-378 tion of position enables the simulation to be performed time-379 independently by simulating each neutron wavelength sepa-380 rately, which greatly reduces computation time.

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The Bloch-equation based neutron polarization simulation done with a 6th order Runge-Kutta method. In order to 382 is maintain accuracy through the numerical iteration process, it is necessary to decompose the magnetic field component of both the gradient magnetic field and RF-field into proper relatively infinitesimal step size, so that the magnetic field can be treated as constant in each calculation. This condition is achieved by assigning the combination of the gradient and RF-field based on the travel distance along the neutron path, so that the adiabaticity can be cross-checked based on Eq. (2). Neutrons with different wavelength are therefore treated separately with a 0.1Å step size because the difference in their travel speeds leads to different distributions of the RF-field, shown in Fig. 4(c)(e)(g). For the purpose of simplification, all simulations of the polarization transfer start with the neutron polarization vector parallel to the magnetic field outside the main RF-flipping region, so that the effect of the RF-flipping process is estimated independently. The relative phase between the RF signal and the neutron pulse are set to zero across all simulations under the condition that adiabaticity is well above 10 as defined by Eq. (2). It should be noted that while the neutron pulse and the RF signal are not synchronized, the adiabatic process within the rotating frame guarantees that the variation in phase does not cause a difference in the final polarization results. For processes that involve insufficient adiabaticity, results are averaged across all phases to simulate the actual experimental process, where signals from cumulative measurements are collected together by the neutron detector.

The simulation results of the composite magnetic field for 411 neutrons of different wavelengths are shown in Fig. 4. In 412 Fig. 4(c), (e), and (g), the RF-field experienced along the 413 flight path varies depending on the neutron velocity, while 414 the positional gradient field is identical for all neutron wave-415 lengths. Under these simulation conditions, the changing 416 trends of the polarization vectors in the x-direction and y-The polarization vector in the xdirection are similar. direction is not shown in Fig. 4(d), (f), and (h), allowing for a more clear observation of the changing process of the polar-420 ization vector at different positions.

422 flipper, the polarization vector in the z-direction gradually de-477 [39, 42]. With a maximum lifetime exceeding 200 hours, the trons to shift towards the xy-plane. After passing through the 480 imental requirements. The detector used in the measurement 426 center of the RF-field, the polarization vector of the neutron 481 is a standard ³He tube detector with time-of-flight measure-

370 length. Therefore, the longer the neutron wavelength, the 427 undergoes a shift towards the negative z-direction, but at a

For neutrons that satisfy the adiabatic condition, the ac-433 tion of the RF-field changes the polarization vector from the z-direction to the xy-plane and then to the negative z-direction, resulting in a spin-flip. The changes in the polar-436 ization vectors of 4 Å and 8 Å neutrons are as expected, and 437 successful spin flipping has been achieved. Due to their lower velocity, the 8 Å neutrons exhibit more precession cycles dur-439 ing the flipping process.

The dependency of the flipping efficiency (f) on the neu-441 tron wavelength is further simulated across the wavelength 442 range of 1.0–5.5 Å with a step size of 0.1 Å, shown as black 443 dots in Fig. 5. It can be seen from Fig. 5 that the simulated 444 flipping efficiency increases with the neutron wavelength. For 445 neutrons with a wavelength of 4 Å or longer, the flipping ef-446 ficiency saturates above 97%, while for neutron wavelengths shorter than 2.5 Å, the flipping efficiency is lower than 90%, which is considered insufficient. In Fig. 5, the simulated results are compared to experimental measurements to verify 450 and illustrate the flipping process.

NEUTRON EXPERIMENT RESULTS

The prototype device is constructed based on the parame-453 ters previously discussed, and its wavelength dependent flipping efficiency is tested on beamline 20 (BL-20) of the CSNS using a wavelength range of 1.0-5.5 Å selected by the neu-456 tron chopper [38–41]. The setup of the experimental components is shown in Fig. 6. A V-shaped supermirror polar-458 izer manufactured by Swiss Neutronics with a tapered angle 459 of 1.19° and a critical wavelength of 2 Å is used to polar-460 ize neutrons. The guide field, composed of Nd₂Fe₁₄B and 461 iron plates, maintains the polarized transmission of neutrons. 462 Due to the internal magnetic field of the adiabatic RF-flipper 463 being oriented oppositely to the direction of the supermirror 464 magnetic field, a 180° rotation of the guiding magnetic field 465 is implemented in front of the flipper to avoid a zero field 466 region and optimized the transmission of polarized neutrons with wavelengths greater than 1 Å. The guide field generated 468 by the wider window of the RF-flipper distributes larger stray 469 fields then the narrower side, which can interfere with the po-470 larized ³He system and further reduce the ³He analyzing abil-471 ity. In order to optimize the experimental measurements and 472 minimize interference, the narrower window of RF-flipper is 473 arranged as the neutron incident side.

The in-house developed in-situ pumped ³He neutron spin 475 filter, utilizing adiabatic fast passage, can manipulate the po-As shown in Fig. 4(d), when 1 Å neutrons enter the RF- 476 larization state of ³He and serve as a neutron spin analyzer creases while the polarization vector in the y-direction grad- 478 system can operate stably for extended periods, and the maxiually increases, causing the polarization vector of the neu- 479 mum saturated ³He polarization exceeds 70%, meeting exper-

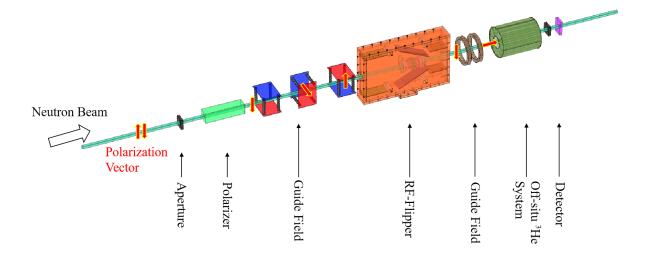


Fig. 6. Schematic diagram of the adiabatic RF-flipper experimental setup on BL-20 of the CSNS.

483 to the power of the CSNS proton pulses to eliminate primary 515 rically, leading to a significant magnetic field gradient at the beam power fluctuations.

flipper, four independent neutron transmission measurements 518 centerline of the RF-flipper, while the experimental results inlel or anti-parallel to the magnetic field. The detection coil 521 the flipping efficiency results in the short wavelength range. measured an RF-field amplitude of 22 G at the center of the solenoid when the device is switched on. The p, a, and fare the efficiencies of the supermirror polarizer, ³He analyzer, and RF-flipper, respectively. In measurements M1 and M2, the RF-flipper is unpowered while the ³He analyzer direction is parallel (for M1) and anti-parallel (for M2) to the magnetic guide field. Measurements M3 and M4 have the same 3 He analyzer state as measurements M1 and M2, respectively, except that the RF-flipper is powered. The normalized intensity of the measured values is determined by

$$M_{1,2} = \frac{1}{4}(1 \pm ap)I_0T_aT_fT_p, \tag{3a}$$

$$M_{3,4} = \frac{1}{4} (1 \pm afp) I_0 T_a T_f T_p.$$
 (3b)

503 ton beam power and T_a , T_f , and T_p are the proportion of neutron transmission through the analyzer, RF-flipper, and polar- 539 son to the centerline simulation results which overestimated 505 izer, respectively. From these quantities, it follows that

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$$f = \left(\frac{M_4 - M_3}{M_4 + M_3}\right) / \left(\frac{M_1 - M_2}{M_1 + M_2}\right) \tag{4}$$

The measured flipping efficiency, shown as red dots in 508 Fig. 5, exceeds 97% for polarized neutrons with a wavelength greater than 4Å. It can be seen that as the neutron wavelength 510 increases, both the simulated and experimental flipping ef-511 ficiency increase. The experimental efficiency of the RF-512 flipper is lower than the simulation results across all wave-513 lengths. The divergence is discovered to be caused by the per-

ment capability. The measured neutron counts are normalized 514 manent magnets of the RF-flipper not being installed symmet-516 beam interface and thereby reducing the flipping efficiency. To obtain the flipping efficiency of the adiabatic RF- 517 Furthermore, the simulated flipping efficiency is based on the are made with a combination of the RF-flipper powered or 519 clude data from the entire neutron beam cross-section transunpowered and the ³He neutron spin filter polarized paral- 520 mitted through the RF-flipper, which leads to discrepancies of

> The difference between the simulation and experiment is 523 further examined by analyzing the irregular gradient field in the effective beam cross-section depicted in Fig. 7(a). A clear 525 variation from the central field of 80 G is observed, ranging 526 from 88 G at the top of the effective neutron beam region to 527 80 G at the center of the solenoid, and ranging from 72.5 G 528 at the far left of the effective neutron beam region to 80 G 529 at the center of the solenoid. Thus, in the experiment as the 530 neutron beam traverses regions away from the center of the 531 solenoid, the non-uniform gradient field leads to a reduction in the flipping efficiency.

The discrepancy is estimated by averaging the polarization transfer across eight locations, 2.5 cm from center in the beam (3b) 535 region, and the results are shown in Fig. 7(b). The green 536 and black simulated results utilize the wavelength-dependent where I_0 is the neutron beam intensity normalized to the pro- 537 RF-field profiles plotted in Fig. 4(c)(e)(g), and the position-538 dependent gradient field plotted in Fig. 7(a). In compari-540 the flipping efficiency, the off-center simulation results are much lower than the experimental results. This result shows (4) 542 that the entire gradient field region traversed by the neutron 543 beam must be considered to create a more accurate simula-544 tion. Multiple averages and calculations are done for the to-545 tal cross section to compare to the experimental results. It 546 has been observed that the magnetic field configuration can 547 be easily influenced by the surrounding environment, which 548 leads to variations in the final flipping efficiency.

> The simulation results also provide the adiabatic parameters of polarized neutrons at different positions: $\kappa = \frac{B\gamma_n}{\frac{d\theta}{d}} \cdot \frac{m\lambda}{h}$,

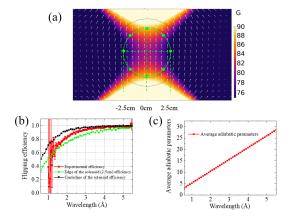


Fig. 7. (a) Cross-sectional view at the center of the solenoid. The 569 white arrows are the directions of the magnetic field lines. The eight green dots are symmetrically distributed on a circle with a radius of 2.5 cm from the center of the solenoid. (b) The green points are the simulated average flipping efficiency of neutrons traversing through the locations of the green dots in (a). The black and red points are the simulated flipping efficiency along the centerline of the solenoid and the experimental flipping efficiency, respectively, as previously shown in Fig. 5. (c) Average adiabatic parameter of different wavelengths.

where κ represents the adiabatic parameter, θ is the angle of 582 the adjacent magnetic field variations vector, l is the distance 583 now be designed in-house while satisfying various beamline $_{553}$ the neutron travels and B is the magnetic field strength. Af- $_{584}$ conditions. Future designs can be improved by optimizing the 554 ter transforming the magnetic field in Fig. 4 from the labo-585 gradient field design to be more uniform throughout the entire 555 ratory frame to the rotating frame, the average adiabatic pa-586 transverse cross-section of the neutron beam path through the 556 rameter for neutrons ranging from 0.6 Å to 5.5 Å varies with 587 RF-field region and improving the RLC-circuit to allow for a 557 wavelength, as shown in Fig. 7(c). It is observed that as the 588 stronger power output.

558 wavelength increases, the average adiabatic parameter of po-559 larized neutrons exhibits a linear upward trend, which aligns with the theoretical principles. However, even with an average adiabatic parameter of 10 for 2 Å neutrons, which meets the adiabatic condition, simulation results indicate a flipping efficiency of only 90% for these neutrons. The investigation implies that phase difference in the simulation does not cause the dropping of the flipping efficiency, and the numerical simulation remains valid for the purpose of understanding the flipping process.

CONCLUSION AND DISCUSSION

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We have successfully designed and manufactured an adi-570 abatic RF-flipper prototype at the China Spallation Neutron 571 Source based on the principles of RF-flipper operation, and 572 the parameters have been optimized for our application. By 573 analyzing the neutron wavelength and flipping efficiency de-574 pendence with simulations and computer-aided design tech-575 niques, precise designs of adiabatic RF-flippers for specific environments can be achieved. The prototype is tested at BL-20 over a wavelength range of 0.6 Å to 5.5 Å, and a flipping 578 efficiency of 97% is achieved for neutrons with wavelength ₅₇₉ above 4 Å. The discrepancy between the design simulations and experiment is caused by an irregular gradient field, and its effect has been investigated through follow-up simulations.

For future applications, an adiabatic RF-flipper device can

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